



# ACE

## Engineering Academy



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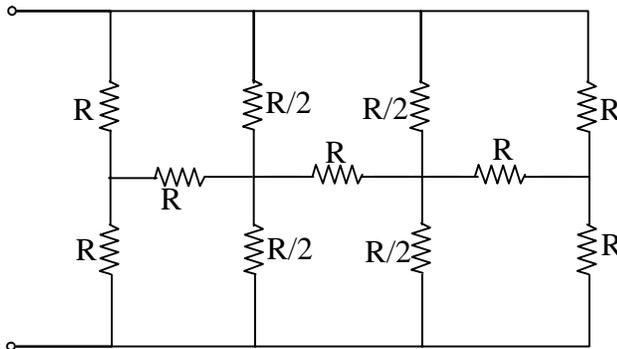
### GATE - 2015 - Electronics & Communication Engineering (EC)

### SOLUTIONS

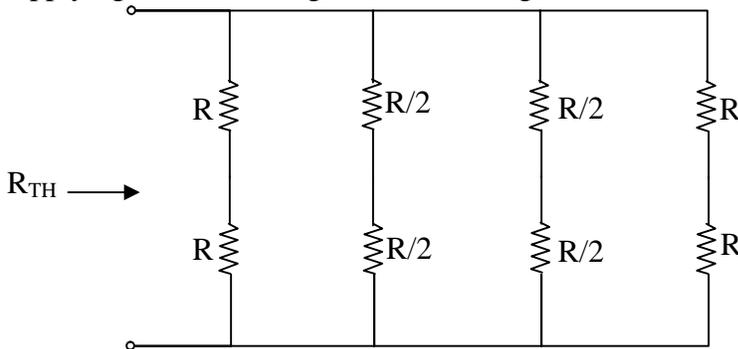
### SET - 1 (31<sup>st</sup> January Forenoon Session)

01. Ans: 100

Sol:



Applying balanced bridge condition we get



$$R_{TH} = \frac{R}{3} = \frac{300}{3} \Rightarrow R_{TH} = 100 \Omega$$

02. Ans: (25)

Sol: At resonant frequency the magnitude of voltage across capacitor is

$$\Rightarrow V_c = QV \angle -90^\circ$$

$$\omega_o = \frac{1}{\sqrt{LC}} = \frac{1}{\sqrt{0.1 \times 10^{-3} \times 1 \times 10^{-6}}} = 1 \times 10^5 \text{ rad/s}; Q = \frac{\omega_o L}{R}$$

$$V_c = 2.5 \times 10 \angle -90^\circ = 25V$$



**03.Ans: (b)**

**Sol:** The characteristic equation for series RLC circuit is

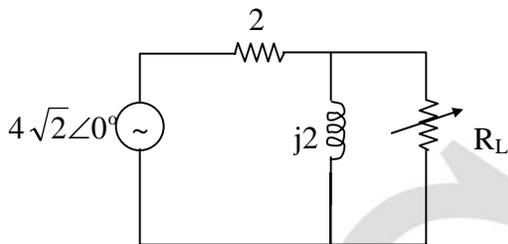
$$s^2 + \frac{R}{L}s + \frac{1}{LC} = 0 \text{ ----- (1)}$$

Standard characteristic equation for any 2<sup>nd</sup> order system is  $s^2 + 2\xi\omega_n s + \omega_n^2 = 0$  ----- (2)

Comparing (1) & (2) we get  $\xi = \frac{R}{2} \sqrt{\frac{C}{L}}$

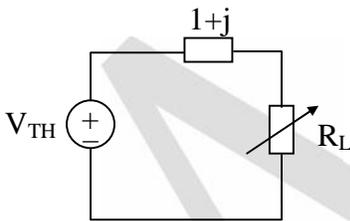
**04.Ans: 1.656**

**Sol:**  $V_m = \sqrt{2} V_{\text{rms}} = 4\sqrt{2} \text{ V}$



$$Z_{\text{TH}} = j2 // 2 = \frac{2 \times j2}{2 + j2} = 1 + j$$

$$V_{\text{TH}} = \frac{j2 \times 4\sqrt{2}}{2 + j2} = 4 \angle 45^\circ$$



$$R_L = \sqrt{1^2 + 1^2}$$

$$R_L = \sqrt{2} \Omega$$

$$V_{R_L} = \frac{R_L V_{\text{TH}}}{1 + j + R_L} = \frac{(4 \angle 45^\circ)(\sqrt{2})}{1 + \sqrt{2} + j}$$

$$V_{R_L} = 2.1647 \angle 22.5^\circ$$

$$I_{R_L} = 1.53 \angle 22.5^\circ$$

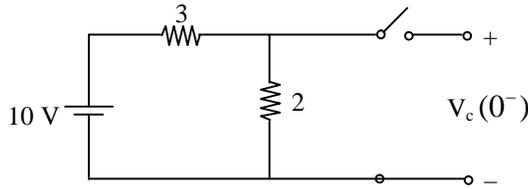
$$P_{R_L} = V_{\text{rms}} I_{\text{rms}} \cos(\theta - \phi)$$

$$= \frac{2.1647 \times 1.53}{2} \cos 0 = 1.6567 \text{ W}$$



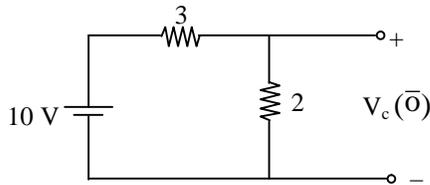
05. **Ans: 2.528V**

**Sol:** The circuit at  $t = 0$  is



$$V_C(0) = V_C(0^-) = V_C(0^+) = 0V$$

The circuit at  $t = \infty$



$$V_C(\infty) = \frac{2}{2+3} \times 10 = 4V$$

$$\tau = RC = 3/2 \text{ (5/6)}$$

$$\tau = 1 \text{ sec}$$

$$V_C(t) = V_C(\infty) + [V_C(0) - V_C(\infty)] e^{-t/\tau}$$

$$\therefore V_C(t) = 4 + [0 - 4]e^{-t}$$

$$V_C(t) = 4(1 - e^{-t}) \text{ V}$$

At  $t = 1 \text{ sec}$

$$\therefore V_C(1) = 4(1 - e^{-1}) = 2.528 \text{ V}$$

$$V_C(1) = 2.528 \text{ V.}$$

06. **Ans: (c)**

**Sol:**  $\therefore V_C(\infty) = 3 \text{ V}$  (i.e) capacitor

Would have reached final steady state value, in this process 50% of energy is also dissipated in the resistor as heat while changing capacitor.

$$E_{\text{absorbed}} = E_{\text{stored (capacitor)}} + E_{\text{dissipated (resistor)}}$$

$$V_c(t) = V_s [1 - e^{-t/\tau}] \dots\dots\dots (1)$$

$$i_c(t) = \frac{CdV_c}{dt} = CV_s \left[ \frac{1}{RC} \right] e^{-t/\tau}$$

$$i_c(t) = \frac{V_s}{R} e^{-t/\tau} \dots\dots\dots (2)$$

$$P_c(t) = V_c(t) \cdot i_c(t)$$

$$P_c(t) = \frac{V_s^2}{R} [e^{-t/\tau} - e^{-2t/\tau}] \dots\dots\dots (3)$$

$$E_c(t) = \int_0^{\infty} p_c(t) dt$$

Energy stored upto  $t =$   
is energy at  $t =$



$$= \frac{1}{2} [V_s]^2$$

$$= \frac{1}{2} C V_s^2 = 0.45 \text{ J}; i_R(t) = i_C(t)$$

$$i_R(t) = \frac{V_s}{R} e^{-t/\tau}$$

$$V_R(t) = V_s e^{-t/\tau}$$

$$P_R(t) = \frac{V_s^2}{R} e^{-2t/\tau}$$

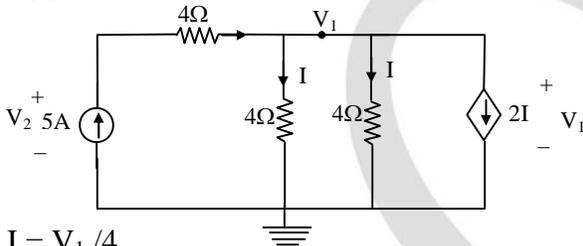
$$E_R(t) = \int_0^{\infty} P_R(t) dt$$

$$= \frac{1}{2} C V_s^2 = 0.45 \text{ J}$$

So, total energy taken from source = 0.9 J

07. Ans: (a)

Sol:



$$I = V_1 / 4$$

Applying KCL at node A we have

$$-5 + I + I + 2I = 0$$

$$4I = 5$$

$$I = \frac{5}{4} \text{ A}$$

$$V_1 = 4I = 5 \text{ V}$$

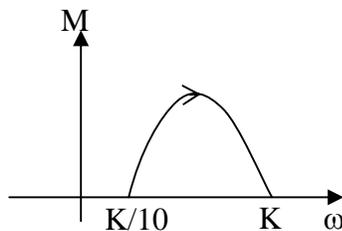
$$V_2 - 4 \times 5 = 5 \text{ V} \Rightarrow V_2 = 25 \text{ V}$$

08. Ans: (a) I<sup>st</sup> Quadrant

$$\text{Sol: } M = \frac{K\sqrt{\omega^2 + 1}}{\sqrt{\omega^2 + 100}}$$

$$\phi = \tan^{-1} \omega - \tan^{-1} \frac{\omega}{10}$$

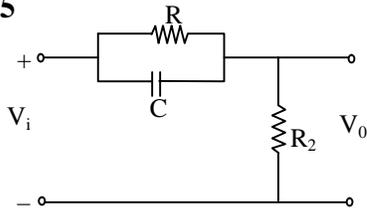
$\omega$	M	$\phi$
0	K/10	0
10	0.7106K	39.29
$\infty$	K	0





**09. Ans: 0.5**

**Sol:**



Standard TF of a lead n/w is

$$G(S) = \alpha \frac{(1 + s\tau)}{(1 + \alpha s\tau)} \quad (1)$$

Given TF function is  $\frac{s + 2}{s + 4}$  — (2)

Comparing (1) & (2)

$$\tau = 0.5 ; \alpha\tau = 0.25$$

$$\Rightarrow \alpha = 0.5$$

$$\therefore RC = \tau = 0.5$$

**10. Ans : (a)**

**Sol :**  $G(s)H(s) = \left( K_p + \frac{K_I}{s} \right) \left( \frac{1}{s(s+5)} \right)$

$$G(s)H(s) = \frac{(sK_p + K_I)}{s^2(s+5)}$$

CE  $\Rightarrow 1 + G(s)H(s) = 0$

$$s^3 + 5s^2 + sK_p + K_I = 0$$

$s^3$	1	$K_p$
$s^2$	5	$K_I$

$s^1$	$\frac{5K_p - K_I}{5}$
-------	------------------------

$s^0$	$K_I$
-------	-------

for stability  $K_I > 0$        $5K_p > K_I$

$$\therefore 5K_p > K_I > 0$$

**11. Ans : (b)**

**Sol :** Negative feedback in a closed loop control system reduces the overall gain of the system. As the gain-bandwidth product of any system should be constant, so the bandwidth should increase. Negative feedback improves disturbance rejection and it reduces the sensitivity to parameter variation.

**12. Ans: K = 25.54**

**Sol:**  $G(s)H(s) = \frac{K(s+4)}{(s+8)(s^2-9)}$

Any point lies on the root locus if it satisfies the angle criteria. The point satisfying angle criteria definitely satisfies magnitude criteria.

Angle criteria  $\angle G(s)H(s) = (2q + 1)180^\circ$

Magnitude criteria  $|G(s)H(s)| = 1$



$$\angle G(s)H(s) = \{\angle(-1 + j2 + 4) - [\angle(-1 + j2 + 8) + \angle(-1 + 2j)^2 - 9]\}$$

$$\approx 180^\circ$$

∴ the point satisfies angle criteria

$$\therefore \left| \frac{K(-1 + j2 + 4)}{(-1 + j2 + 8)[(-1 + j2)^2 - 9]} \right| = 1$$

$$\left| \frac{K(\sqrt{13})}{\sqrt{53}\sqrt{60}} \right| = 1$$

$$\Rightarrow K = 25.54$$

**13. Ans: 12**

**Sol:** Characteristic equation is

$$S^3 + 4S^2 + 3S + K = 0$$

By Routh criteria condition for marginal stability is  $k = 12$ .

**14. Ans: 0.52**

**Sol:**  $\mu_n = 1200 \text{ cm}^2 / \text{Vs}$  ;

$$\mu_p = 400 \text{ cm}^2 / \text{Vs} ;$$

$$N_D = 1 \times 10^{16} / \text{cm}^3$$

$$\sigma = n_n q \mu_n + p_n q \mu_p$$

For Si  $n_i = 1.5 \times 10^{10}$  ;  $n_n = N_D = 1 \times 10^{16} / \text{cm}^3$

$$p_n = \frac{n_i^2}{N_D} = \frac{(1.5 \times 10^{10})^2}{1 \times 10^{16}} = 22500$$

As  $n_n \gg p_n$

$$\therefore \sigma = N_D q \mu_n = (1 \times 10^{16})(1.6 \times 10^{-19})(1200)$$

$$= 1.92 (\Omega\text{-cm})^{-1}$$

$$\rho = 0.52 \Omega\text{-cm}$$

**15. Ans: (a)**

**Sol:** Negative differential resistance is exhibited in P-N junction diode which are heavily doped on both sides

Eg: Tunnel diode.

**16. Ans: 2.5**

**Sol:**  $C_J = \frac{C_{J0}}{\left(1 + \frac{V_R}{V_o}\right)^m}$  ;  $m = \frac{1}{2}$  for abrupt junction.



$$\frac{C_{J1}}{C_{J2}} = \frac{\left(1 + \frac{V_{R2}}{V_o}\right)^{1/2}}{\left(1 + \frac{V_{R1}}{V_o}\right)^m} = \frac{\left(1 + \frac{7.25}{0.75}\right)^{1/2}}{\left(1 + \frac{1.25}{0.75}\right)^{1/2}}$$

$$\frac{C_{J1}}{C_{J2}} = 2 \Rightarrow C_{J2} = \frac{5}{2} = 2.5$$

$$\Rightarrow C_{J2} = 2.5\text{pF}$$

**18. Ans: (a)**

**Sol:**  $V_D = V_G = V_{DD}$

$$V_{DS} = V_{GS} = V_{DD} - V_{SS}$$

$$I_D = K[V_{GS} - V_T]^2$$

$$I_D = K[V_G - V_{SS} - V_T]^2$$

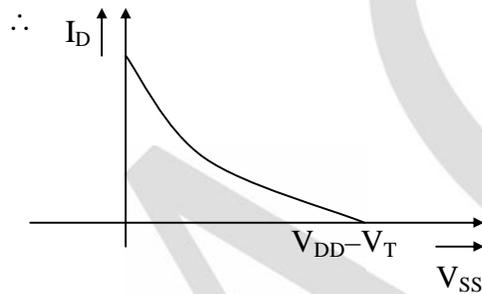
$$I_D = K_1[(V_{DD} - V_T) - V_{SS}]^2$$

$$\frac{\partial I_D}{\partial V_{SS}} = 2K_1[V_{DD} - V_T - V_{SS}]$$

$$= 2K_1[K_2 - V_{SS}]$$

$\Rightarrow$  Slope of the curve is negative

At  $V_{SS} = V_{DS} - V_T$ ;  $I_{DS} = 0$



**19. Ans: 20.50 kh**

**Sol:** Given  $\lambda = 0.05\text{V}^{-1}$

$$I_D = 1\text{mA}; V_{DS} = 0.5\text{V}$$

$$I_D = k[V_{GS} - V_T]^2 (1 + \lambda V_{DS}) \text{ ----- (1)}$$

$$k[V_{GS} - V_T]^2 = \frac{I_D}{1 + \lambda V_{DS}}$$

$$= \frac{1 \times 10^{-3}}{1 + 0.05 \times 0.5}$$

$$k[V_{GS} - V_T]^2 = \frac{1}{1025} \text{ ----- (2)}$$

output resistance  $r_o = \left. \frac{\partial V_{DS}}{\partial I_D} \right|_{V_{GS} \rightarrow \text{const}}$

Differentiating (1) we get

$$\frac{\partial I_D}{\partial V_{DS}} = k[V_{GS} - V_T]^2 \lambda \text{ ----- (3)}$$



Substituting (2) in (3) we get

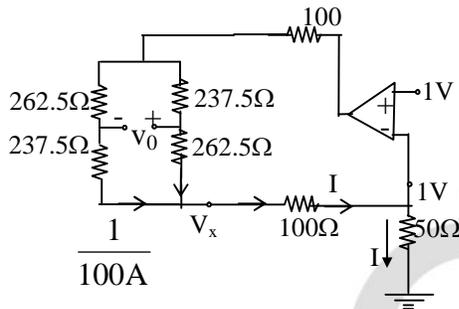
$$\frac{\partial I_D}{\partial V_{DS}} = \frac{1}{1025} \times 0.05$$

$$\frac{\partial V_{DS}}{\partial I_D} = r_o = \frac{1}{4.878 \times 10^{-5}}$$

$$r_o = 20.50 \text{ k}\Omega$$

**20. Ans: 250**

**Sol:**



By virtual ground concept  $V_i^+ = V_i^- = 1V$

$$I = \frac{1}{50} \text{ A}$$

$$V_x = 100 I + 1$$

$$= \frac{100}{50} + 1$$

$$= 3 \text{ V}$$

$$V_o^+ - 262.5 \times \frac{1}{100} = 3$$

$$V_o^+ = 5.625 \text{ V}$$

$$V_o^- = \frac{237.5}{100} + 3V = 5.375 \text{ V}$$

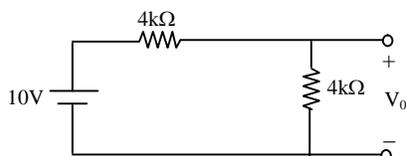
$$V_o = V_o^+ - V_o^-$$

$$V_o = 0.25 \text{ V}$$

$$V_o = 250 \text{ mV}$$

**21. Ans: 5**

**Sol:** The region of operation of the zener diode should be determined first. Removing the zener diode from the circuit above we have the circuit as





$V_0 = \frac{4}{4+4} \times 10 = 5V$  as the voltage across the zener is less than 6 V so the zener diode does not operate in reverse break down region.  
 $\therefore V_0 = 5V$

**22. Ans : 83.85**

**Sol :**  $I_D = I_S e^{\frac{V_{BE}}{V_t}}$

$$V_{BE} = V_t \ln \left[ \frac{I_D}{I_S} \right]$$

From the given figure

$$\begin{aligned} V_{12} &= V_1 - V_2 \\ &= V_t \ln \left[ \frac{I_{D1}}{I_S} \right] - V_t \ln \left[ \frac{I_{D2}}{I_S} \right] \\ &= V_t \left[ \ln \left( \frac{I_{D1}}{I_{D2}} \right) \right] \text{-----(1)} \end{aligned}$$

It is known that

$$\begin{aligned} V_t &= \frac{T}{11600} \\ \frac{V_{t_1}}{T_1} &= \frac{V_{t_2}}{T_2} \\ V_{t_2} &= \frac{T_2}{T_1} \times V_{t_1} \end{aligned}$$

$$= \frac{323}{300} (26m)$$

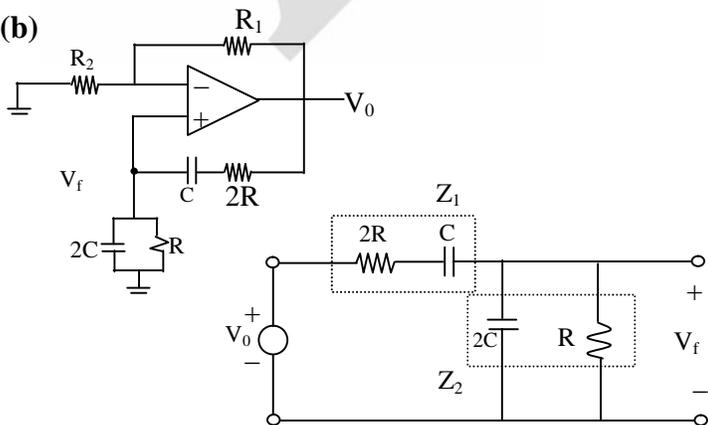
$$V_{t_2} = 27.99 \text{ mV} \text{ -----(2)}$$

Substitute the above value in equation (1)

$$\therefore V_{12} = 27.99 \text{ m} \times \ln \left( \frac{80m}{4m} \right) \Rightarrow V_{12} = 83.85 \text{ mV.}$$

**23. Ans : (b)**

**Sol :**





$$Z_1 = 2R + \frac{1}{sC}$$

$$= \frac{1 + 2sRC}{sC}$$

$$Z_2 = \frac{R \times \frac{1}{2sC}}{R + \frac{1}{2sC}} = \frac{R}{2sRC + 1}$$

$$= \frac{R}{R + \frac{1}{2sC}}$$

$$\therefore V_f = V_0 \times \frac{Z_2}{Z_1 + Z_2}$$

$$= V_0 \times \frac{\frac{R}{1 + 2sRC}}{\frac{1 + 2sRC}{sC} + \frac{R}{1 + 2sRC}}$$

$$= V_0 \frac{R}{(1 + 2sRC)} \times \frac{sC(1 + 2sRC)}{(1 + 2sRC)^2 + sRC}$$

$$= \frac{V_0 RCS}{1 + 4sRC + 4s^2R^2C^2 + sRC}$$

$$V_f = \frac{V_0 RCS}{1 + 5RCS + 4R^2C^2s^2}$$

$$= \frac{V_f}{V_0} = \frac{RSC}{1 + 5RCS + 4R^2C^2s^2}$$

$$\text{Gain } A = 1 + \frac{R_1}{R_2}$$

For oscillations to occur

$$A = 1$$

$$\Rightarrow A = \frac{1}{\beta}$$

$$1 + \frac{R_1}{R_2} = \frac{1 + 5RCS + 4R^2C^2s^2}{sRC}$$

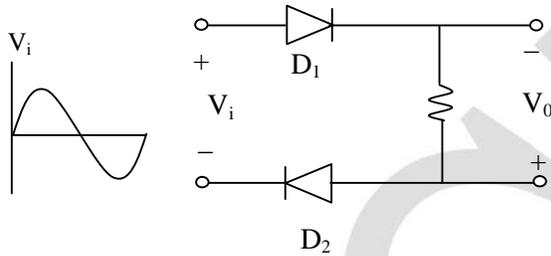


Equating Real & Imaginary parts

$$\begin{array}{l|l}
 1 + \frac{R_1}{R_2} = 5 & \frac{1}{RCs} + 4RCs = 0 \\
 \frac{R_1}{R_2} = 4 & 4R^2s^2C^2 + 1 = 0 \\
 R_1 = 4R_2 & \omega_o = \frac{1}{2RC}
 \end{array}$$

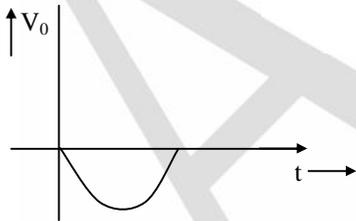
**24. Ans : (c)**

**Sol:** The given circuit diagram can be redrawn as



(a) For +ve half cycle :  
For +ve half cycle the diodes  $D_1$  and  $D_2$  are ON  
 $V_0 = -V_{in}$

(b) For -ve half cycle :  
For -ve half cycle the diodes  $D_1$  and  $D_2$  are off



**25. Ans: (b)**

**Sol:** The result of addition operation is stored in accumulator (A) and the status of carry bit is stored in the PSW i.e flag register (F).

**26. Ans: (a)**

**Sol:**  $F = xy\bar{z} + x\bar{y}z + x\bar{y}\bar{z} + \bar{x}y\bar{z}$

	00	01	11	10
0	0	0	0	1
1	1	1	0	1

$$F = (x+y+z)(x+y+\bar{z})(x+\bar{y}+\bar{z})(\bar{x}+\bar{y}+\bar{z})$$



27. Ans: (a)

Sol :  $M(a,b,c) = ab+bc+ca$

K map for  $M(a,b,c)$

		bc			
a		00	01	11	10
0		0	0	1	0
1		0	1	1	1

Minterms are 3, 5, 6, 7

K map for  $M(a,b,c)$

		bc			
a		00	01	11	10
0		1	1	0	1
1		1	0	0	0

Minterms are 0, 1, 2, 4

$$m(a, b, c) = \bar{b}\bar{c} + \bar{a}\bar{b} + \bar{a}\bar{c} = x$$

$$C = Z$$

$$M(X, Y, Z) = XY + YZ + ZX$$

$$= (\bar{b}\bar{c} + \bar{a}\bar{b} + \bar{a}\bar{c})(ab + b\bar{c} + a\bar{c}) + (\bar{b}\bar{c} + \bar{a}\bar{b} + \bar{a}\bar{c})(c)$$

$$= a\bar{b}\bar{c} + \bar{a}b\bar{c} + abc + \bar{a}\bar{b}c$$

$$= \bar{c}[a\bar{b} + \bar{a}b] + c[ab + \bar{a}\bar{b}]$$

$$= \bar{q}p + q\bar{p}$$

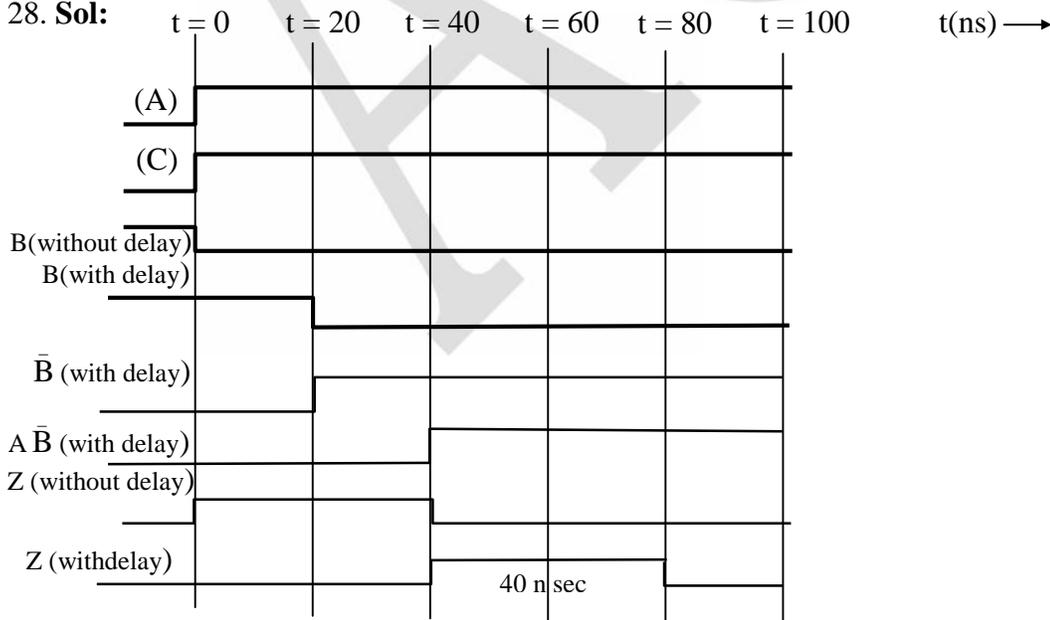
$$= p \oplus q ; \text{ where } q = c, p = a \oplus b$$

$$\therefore M(x, y, z) = a \oplus b \oplus c$$

$\therefore$  3 input X-OR realizes the equation

$$M[\overline{M(a, b, c)}, M(a, b, \bar{c}), c]$$

28. Sol:



$\therefore$  Z is 1 for 40 nsec



**29. Ans: 7**

**Sol:** 16kb memory is to be designed with no of rows = no. of columns

Let no of rows = no. of columns = x.

$$16 \times 1024 = 16384 = x^2$$

$$\Rightarrow x = 128$$

i.e. 128 rows and 128 columns.

To address 128 row let no. of decoder address lines = n.

$$2^n = 128 \Rightarrow n = 7$$

**30. Ans: 0.937 V**

**Sol:** 0001  $\rightarrow$  0.0625 V corresponds to step size.

$$\therefore f_{so} = (2^n - 1) \times \text{step size}$$

$$= 15 \times 0.0625$$

$$= 0.9375$$

**31. Ans: 8.885**

$$\text{Sol: } \vec{E} = 2 \cos\left(10^8 t - \frac{Z}{\sqrt{2}}\right)$$

$$\beta = \frac{1}{\sqrt{2}} \text{ rad/m}; \omega = 10^8 \text{ rad/s}$$

$$\lambda = \frac{2\pi}{\beta} = \frac{2\pi}{\frac{1}{\sqrt{2}}} = 8.885\text{m}$$

**32. Ans: (a)**

$$\text{Sol: } H_\phi = \frac{I}{2\pi r} \hat{a}_\phi$$

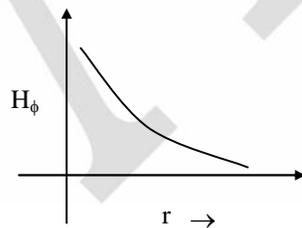
r = radial distance

I = current in the filament

$$H_\phi = \frac{I}{2\pi} \left(\frac{1}{r}\right) \hat{a}_\phi$$

$$H_\phi = K \left(\frac{1}{r}\right)$$

$$H_\phi \propto \left(\frac{1}{r}\right)$$





33. Ans: (c)

Sol:  $H_z = \cos m \frac{\pi}{a} x \cos n \frac{\pi}{b} y \cos \beta z$

$$\frac{m}{0.08} = 25 \Rightarrow m = 2$$

$$\frac{n}{0.033} = 30.3 \Rightarrow n = 1$$

$\therefore H_z$  is existing so TE<sub>21</sub>

34. Ans: 53.31

Sol:  $E = 24\pi \cos(\omega t - \beta x) \text{ v/m } \hat{a}_z$

$$P_{\text{avg}} = \oint \vec{\delta} \cdot d\vec{s}$$

Surface is given as  $x+y=1$

$$\therefore d\vec{s} = \frac{\hat{a}_x + \hat{a}_y}{\sqrt{2}}$$

Poynting vector

$$\vec{\delta} = \frac{E_0^2}{2\eta} \hat{a}_x$$

$$\therefore P_{\text{avg}} = \frac{(24\pi)^2}{2 \times 120\pi} \times \frac{1}{\sqrt{2}} \times 10 \times 10 \times 10^{-4}$$

$$= 53.31 \text{ mW}$$

36. Ans:(d)

Sol:  $x(t) * h(t) = y(t)$

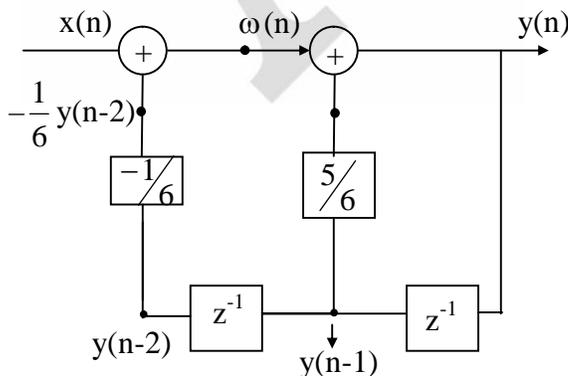
$$x(-t) * h(-t) = y(-t)$$

$$x(t) * \delta(t - t_0) = x(t - t_0)$$

$$x(-t) * \delta(-t - t_0) = x(-t - t_0)$$

38. Ans:(c)

Sol:





$$\omega(n) = x(n) - \frac{1}{6}y(n-1) \text{ ----- (1)}$$

$$y(n) = \omega(n) - \frac{5}{6}y(n-1) \therefore y(n) = x(n) - \frac{1}{6}y(n-2) + \frac{5}{6}y(n-1)$$

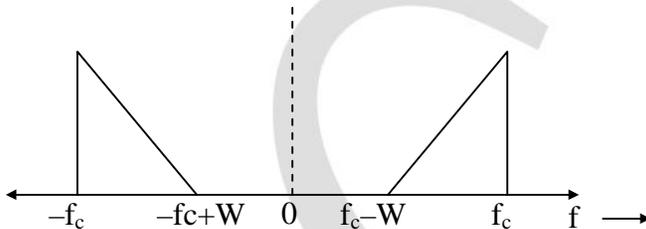
$$\frac{Y(z)}{X(z)} = H(z) = \frac{1}{1 + \frac{1}{6}z^{-2} - \frac{5}{6}z^{-1}} = \frac{1}{\left(1 - \frac{1}{2}z^{-1}\right)\left(1 - \frac{1}{3}z^{-1}\right)}$$

$$\therefore \text{poles at } z = \frac{1}{3}, \frac{1}{2}$$

**41. Ans: (c)**

**Sol:**  $x(t) = m(t) \cos\omega_c t + \hat{m}(t) \sin\omega_c t.$

The given signal is the lower sideband of a SSB – SC.



The signal is clearly a bandpass signal.

**43. Ans: (a)**

**Sol:**  $\Delta = \text{step size} = 0.1 \text{ V}$

$f_s = \text{sampling rate} = 20,000 \text{ samples/sec}$

$f_m = \text{message signal} = 2000 \text{ Hz}$

$A_m$  maximum amplitude of message signal to avoid slope overload distortion

$$\frac{\Delta}{T_s} > 2\pi f_m A_m ;$$

$$0.1 \times 20,000 > 2\pi \times 2000 \times A_m \left( \because f_s = \frac{1}{T_s} \right)$$

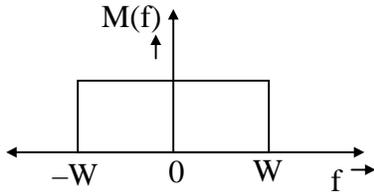
$$A_m < \frac{0.1 \times 20000}{2\pi \times 2000}$$

$$A_m = \frac{1}{2\pi}$$



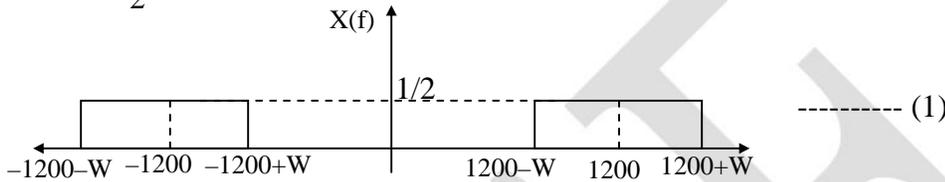
**44. Ans: 350**

**Sol:** Let  $M(f)$  be as shown in figure



$$x(t) = m(t) \cos(2400 \pi t)$$

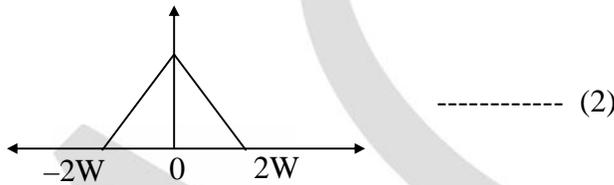
$$X(f) = \frac{1}{2} [M(f - f_c) + M(f + f_c)]; \quad f_c = 1200 \text{ Hz}$$



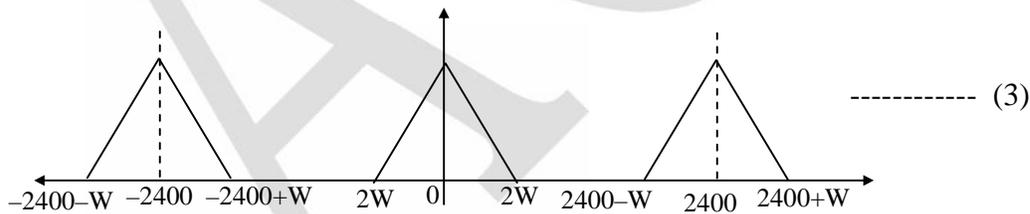
$$Y(f) = X(f) * X(f) + 10 X(f)$$

$$x^2(t) = \frac{m^2(t)}{2} [1 + \cos 4\pi f_c t]$$

Spectrum for  $m^2(t)$  is



Spectrum for  $m^2(t) [1 + \cos 4\pi f_c t]$  is



For the output to be  $z(t) = 10x(t)$

From diagram (1), (2), (3) we have

$$1200 - W \geq 700 \quad \& \quad 1200 + W \leq 1700$$

$$W \leq 500$$

$$\Rightarrow W > 300 \text{ Hz}$$

$$2W \leq 700$$

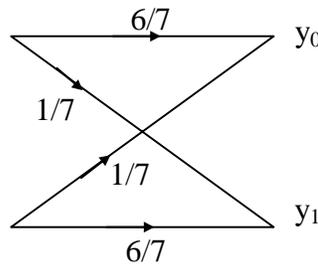
$$W \leq 350$$

$$W_{\max} = 350 \text{ Hz}$$



**46 .Ans: 0.4**

**Sol:**  $p(x_0) = 0.2$   
 $p(x_1) = 0.8$   
 $p(y_0/x_0) = p(y_1/x_1) = 6/7$   
 $p(y_0/x_1) = p(y_1/x_0) = 1/7$   
 $p(x_1/y_0) = ?$



$$p(x_1/y_0) = \frac{p(y_0/x_1)p(x_1)}{p(y_0)} = \frac{p(y_0/x_1)p(x_1)}{p(y_0/x_1)p(x_1) + p(y_0/x_0)p(x_0)} = \frac{\frac{1}{7} \times 0.8}{\frac{1}{7} \times 0.8 + \frac{6}{7} \times 0.2}$$

$p(x_1/y_0) = 0.4$

47. The solution of the differential equation.

$$\frac{d^2x}{dt^2} + 2\frac{dx}{dt} + x = 0, \quad x(0) = x'(0) = 1$$

**Sol :** Auxiliary equation  $D^2+2D+1=0$

$$(D+1)^2 = 0$$

$$D = -1, -1$$

The solution is  $x = (C_1+C_2t)e^{-t}$  -----(1)

$$x' = -(C_1+C_2t)e^{-t} + C_2e^{-t}$$
 -----(2)

$t=0, x=1, (1) \Rightarrow 1 = C_1$

$t=0, x'=1, (2) \Rightarrow 1 = -C_1+C_2 \Rightarrow C_2 = 2$

solution is  $x(t) = (1+2t) e^{-t}$

**48.Ans: 17**

**Sol:**  $AX = \lambda X$

$$\begin{bmatrix} 4 & 1 & 2 \\ P & 2 & 1 \\ 14 & -4 & 10 \end{bmatrix} \begin{bmatrix} 1 \\ 2 \\ 3 \end{bmatrix} = \lambda \begin{bmatrix} 1 \\ 2 \\ 3 \end{bmatrix} = \begin{bmatrix} \lambda \\ 2\lambda \\ 3\lambda \end{bmatrix}$$

$$\lambda = 12$$

$$P+4+3 = 2\lambda = 24$$

$$P = 17$$

**49.Ans: (b)**

**Sol:**  $f(x) = 1-x^2+x^3$  in  $[-1, 1]$

$$f'(x) = -2x + 3x^2$$

from the mean value theorem,

$$f'(c) = \frac{f(b) - f(a)}{b - a} = \frac{f(1) - f(-1)}{2}$$

$$\Rightarrow -2C + 3C^2 = \frac{(1-1+1) - (1-1-1)}{2} = 1$$



$$\Rightarrow 3C^2 - 2C - 1 = 0$$

$$C = 1, -\frac{1}{3}$$

$$\therefore C = -\frac{1}{3} \in (-1, 1)$$

$C = 1$  does not lie in  $(-1, 1)$

**50. Ans: (c)**

**Sol:** Given

$$\vec{P} = (x^3y)\hat{a}_x + (-x^2y^2)\hat{a}_y + (-x^2yz)\hat{a}_z$$

$$(i) \operatorname{div} \vec{P} = \Delta \cdot \vec{P}$$

$$= \frac{\partial}{\partial x}(x^3y) + \frac{\partial}{\partial y}(-x^2y^2) + \frac{\partial}{\partial z}(-x^2yz)$$

$$= 3x^2y - 2x^2y + (-x^2y)$$

$$= 0$$

$\therefore \vec{P}$  is a solenoidal vector

$$(ii) \operatorname{curl} \vec{P} = \Delta \times \vec{P}$$

$$= \begin{vmatrix} \hat{a}_x & \hat{a}_y & \hat{a}_z \\ \frac{\partial}{\partial x} & \frac{\partial}{\partial y} & \frac{\partial}{\partial z} \\ x^3y & -x^2y^2 & -x^2yz \end{vmatrix}$$

$$= (-x^2z - 0)\hat{a}_x - (-2xyz - 0)\hat{a}_y + (-2xy^2 - x^3)\hat{a}_z$$

$$0$$

$\therefore \vec{P}$  is a rotational vector

Hence the given vector  $\vec{P}$  is solenoidal but not irrotational

**51. Ans: 1**

**Sol** Given ellipse equation is

$$x^2 + 3y^2 = 1 \text{ -----(1)}$$

let the area of the rectangle inscribed in the ellipse be  $A = 2x \times 2y$

$$\text{Let it be } f(x, y) = 4xy = 4x \cdot \frac{\sqrt{1-x^2}}{2}$$

$$f(x) = 2x\sqrt{1-x^2}$$

$$f'(x) = 0$$

$$\frac{2x \times -2x}{2\sqrt{1-x^2}} + 2\sqrt{1-x^2} = 0$$

$$\frac{-2x^2}{\sqrt{1-x^2}} + 2\sqrt{1-x^2} = 0$$

$$-2x^2 + 2(1-x^2) = 0$$

$$-4x^2 + 2 = 0$$



$$4x^2 = 2$$

$$x^2 = \frac{1}{2} \Rightarrow x = \pm \frac{1}{\sqrt{2}}$$

$$\text{Area} = 4xy = 2x \sqrt{1-x^2} = 2 \frac{1}{\sqrt{2}} \sqrt{1-\frac{1}{2}}$$

$$= 1$$

**52. Ans: (b)**

**Sol :**  $f(x) = e^{-x}(1+x+x^2)$

$f'(x) = -e^{-x}[1+x+x^2] + e^{-x}[2x+1]$

$f'(x) = 0 \Rightarrow 2x+1-1-x-x^2 = 0.$

$\Rightarrow x = 0/1$

$f''(x) = e^{-x}(1+x+x^2) - e^{-x}(1+2x) - e^{-x}(2x+1) + e^{-x}(2)$

$= e^{-x}[1+x+x^2-1-2x-2x-1+2]$

$= e^{-x}(x^2-3x+1)$

At  $x=1$

$= f''(1) = e^{-1}[1-3+1]$

$= -0.3678$

$f''(1) < 0 \Rightarrow$  matrix at  $x = 1$

from the options,

option (b) satisfies the above analysis.

**53. Sol :** (a)  $f(z) = \frac{1}{z^2-1} = \frac{\left(\frac{1}{z}+1\right)}{z-1}$  has a singular point at  $z=1$

$\text{Res}(f(z) : z=1) = \left(\frac{1}{z+1}\right)_{z=1} = \frac{1}{1+1} = \frac{1}{2}$

$\therefore$  option (a) is a correct statement

(b)  $I = \oint_C f(z) dz = \oint_C z^2 dz$

Here  $f(z) = z^2$  is analytic function everywhere

$\therefore \oint_C z^2 dz = 0$  and option (b) is a correct statement

(c)  $f(z) = \bar{z} = x-iy$

$\Rightarrow u + iv = f(z) = x-iy$

$\Rightarrow u = x \text{ and } v = -y$

$\Rightarrow u_x = 1 \quad v = 0$

$u_y = 0 \quad v_y = -1$

$\text{Here } u_x = 1 \quad v_y = -1$

i.e. one of the C.R. equation is not satisfied.

$\therefore f(z) = \bar{z}$  is not analytic function

Hence the given statement that  $f(z) = \bar{z}$  is analytic, is a wrong statement.

**54. Ans:C**

**Sol:** As events A & B are independent

$$\therefore P(A \cap B) = P(A) P(B)$$

$$P(B/A) = \frac{P(A \cap B)}{P(A)} = \frac{P(A)P(B)}{P(A)}$$

$$= P(B)$$

$$P(\bar{A} \cap \bar{B}) = P(\bar{A})P(\bar{B})$$

$$P(A \cup B) = P(A) + P(B) - P(A \cap B)$$

$$= [P(A) + P(B) - P(A) P(B)]$$

**55.Sol:**

The condition for many solutions of the system  $AX = B$  is

$$\rho(A) = \rho(A/B) < n = 3.$$

Or

For many solutions, we can consider  $|A| = 0$ ,

Because, the constant 'k' is in coefficient matrix.

$$\text{Now } \begin{vmatrix} -1 & 2 & -3 \\ 1 & 6 & 12 \\ 2 & -4 & 3k \end{vmatrix} = 0$$

$$\Rightarrow 24k - 48 = 0$$

$$\Rightarrow k = 2.$$

$\therefore$  The system has many solutions for  $k = 2$ .

**General Aptitude:**

**01.Ans: (b)**

**02.Ans: (a)**

**03.Ans:1.**

$$\text{Sol: } 66 \quad 6 = \frac{66 - 6}{66 + 6} = \frac{60}{72} = \frac{5}{6}$$

$$66 \quad 6 = \frac{66 + 6}{66 - 6} = \frac{72}{60} = \frac{6}{5}$$

$$66 \quad 6 \rightarrow (66 \quad 6) = \frac{5}{6} \times \frac{6}{5} = 1$$

**04.Ans:2.72**

$$\text{Sol: } \log_b a = \frac{\log_e a}{\log_e b}$$



$$\therefore \log_x 5/7 = \frac{\log_e 5/7}{\log_e x} = -1/3$$

$$\therefore \log_e x = 1 \Rightarrow x = e$$

$$\Rightarrow x = 2.72$$

**05. Ans(b)**

**Sol:** Total number of small cubes = 27

Total number of faces =  $27 \times 6 = 162$

Each face of big cube shows 9 faces of small cubes

$\therefore$  Total number of faces visible =  $9 \times 6 = 54$

$\therefore$  Total number of invisible faces =  $162 - 54 = 108$

$$\frac{\text{Visible faces}}{\text{Invisible faces}} = \frac{54}{108} = \frac{1}{2}$$

**06. Ans : (b)**

**08. Ans : (b)**

**09. Ans : (a)**